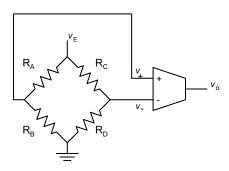
04-1

04-1

04-3

The Wheatstone Bridge



Raison d'être: convert tiny changes in resistance to voltage.

Shown with an instrumentation amplifier.

Like an ideal op-amp but with finite gain.

Gain of instrumentation amplifier denoted by A.

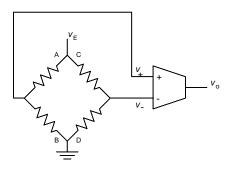
$$v_o = A(v_+ - v_i).$$

The Wheatstone bridge consists of four arms.

$$v_o = A \left(\frac{R_{\rm B}}{R_{\rm A} + R_{\rm B}} - \frac{R_{\rm D}}{R_{\rm C} + R_{\rm D}} \right) v_{\rm E}. \label{eq:vo}$$

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Transducer can be placed in one, two, or four arms.

Typical function: $H_t(x) = R(1+xk), xk \ll 1$

where R is the nominal resistance of the transducer and k is a constant.

For simplicity write function as: $H_t(x) = R + R_s$,

where R is independent of the process-variable value and R_s is dependent on the process-variable value.

Typically, $R \gg R_{\rm s}$.

Usually, need to convert R_s to a voltage.

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04-3

Complementary Pairs

Frequently, transducer pairs can have $complementary\ responses.$

If so, there are two (usually identical) transducers...

- ... positioned so they react oppositely to the process variable...
- \dots so that when their responses are $subtracted\dots$
- ... their response to the process variable add...
- \ldots and unwanted quantities $cancel\ out.$

For example, consider:

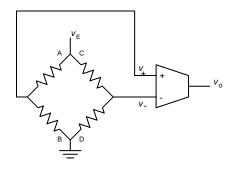
$$H_{t1}(x) = R(1+xk)$$
 and $H_{t2}(x) = R(1-xk)$.

Sum: $H_{t1}(x) + H_{t2}(x) = R$. (Not helpful.)

Difference: $H_{t1}(x) - H_{t2}(x) = 2xk$. (Much better.)

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One-Transducer Configuration



Arm B: $H_t(x) = R + R_s = R(1 + xk)$.

Other Arms: Resistor of value R.

$$v_o = A \left(\frac{R_s}{2(2R + R_s)} \right) v_E \approx A \frac{R_s}{4R} v_E = A \frac{xk}{4} v_E.$$

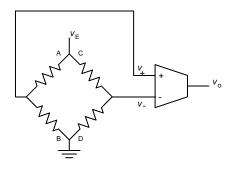
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Two-Transducer Configuration



Arm A: $H_{t2}(x) = R - R_s = R(1 - xk)$.

Arm B: $H_{t1}(x) = R + R_s = R(1 + xk)$.

Other Arms: Resistor of value R.

$$v_o = A \frac{R_s}{2R} v_E = A \frac{xk}{2} v_E.$$

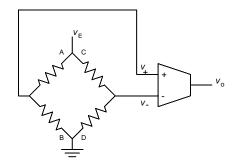
As one might expect, twice as sensitive.

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Four-Transducer Configuration



Arms A and D: $H_{t2}(x) = R - R_s = R(1 - xk)$.

Arms B and C: $H_{t1}(x) = R + R_s = R(1 + xk)$.

$$v_o = A \frac{R_s}{R} v_E = Axk v_E.$$

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Wheatstone Bridge Transfer Functions

Goal

Let $R_{\rm t} = R \pm R_{\rm s} = R(1 \pm xk)$ be the transducer response(s).

Assume bridge designed properly.

Need to find two functions:

$$H_{\rm c}(R_{\rm t}) = \dots$$
 and $H_{\rm c}(R_{\rm s}) = \dots$

Both functions are equivalent.

Choose whichever is more convenient.

Four-Transducer Configuration

$$H_{\rm c}(R_{\rm s}) = v_o = A \frac{R_{\rm s}}{R} v_{\rm E}.$$

Let
$$R_t = R + R_s$$
. Then $R_s = R_t - R$.

$$H_{\rm c}(R_{\rm t}) = A \frac{R_{\rm t} - R}{R} v_{\rm E} = A \left(\frac{R_{\rm t}}{R} - 1\right) v_{\rm E}. \label{eq:hc}$$

Two-Transducer Configuration

$$H_c(R_s) = v_o = \frac{A}{2} \frac{R_s}{R} v_E$$

$$H_{\rm c}(R_{\rm t}) = \frac{A}{2} \frac{R_{\rm t} - R}{R} v_{\rm E} = \frac{A}{2} \left(\frac{R_{\rm t}}{R} - 1 \right) v_{\rm E}. \label{eq:hc}$$

One-Transducer Configuration

$$H_c(R_s) = v_o = \frac{A}{4} \frac{R_s}{R} v_E.$$

$$H_{\rm c}(R_{\rm t}) = \frac{A}{4} \frac{R_{\rm t} - R}{R} v_{\rm E} = \frac{A}{4} \left(\frac{R_{\rm t}}{R} - 1 \right) v_{\rm E}. \label{eq:hc}$$

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Wheatstone Bridge Sample Problem $\begin{aligned} & \textit{Design a system with output } v_o = H(x), \textit{ where process variable } x \textit{ is} \\ & \textit{strain and, } x \in [0, 10^{-5}], \textit{ and } H(x) = 10^6 x \, \text{V}. \end{aligned}$

Strain will be covered in more detail later.

For now, all we need to know is that strain is dimensionless.

Strain is measured by a strain gauge.

Strain gauges frequently used in complementary pairs.

Use strain gauges with response:

$$H_{\rm t}(\epsilon) = R(1 + \epsilon G_f),$$

where ϵ denotes strain and

constant $G_f = 2$.

 $(G_f$ called gauge factor, a dimensionless quantity.)

Position the two strain gauges to obtain response:

$$H_{t}(x) = R(1 + xG_{f})$$
 and $H_{t'}(x) = R(1 - xG_{f}).$

04-8

Derivation of Conditioning Circuit Needed

A Wheatstone bridge is the obvious choice because transducer response is in form $R\pm R_{\rm s}.$

Nevertheless, conditioning-circuit derivation will be presented.

$$H(x) = H_{\rm c}(H_{\rm t}(x))$$

(Analysis performed as though there were one transducer.) $\,$

$$y = H_{\mathrm{t}}(x) = R_{\mathrm{t}} = R(1 + xG_f)$$
. Then $x = \frac{\frac{y}{R} - 1}{G_f}$.

Then
$$H_{\rm c}(y)=H\left(\frac{\frac{y}{R}-1}{G_f}\right)=10^6\frac{\frac{y}{R}-1}{G_f}\,{\rm V}.$$

Response for two-transducer configuration: $H_{\rm c}(R_{\rm t})=rac{A}{2}igg(rac{R_{
m t}}{R}-1igg)v_{
m E}.$

Choose A and $v_{\rm E}$ so that $\frac{A}{2}v_{\rm E}=\frac{10^6}{G_f}\,{\rm V}$ is satisfied.

For example, $v_{\rm E}=10\,{\rm V}$ and $A=10^5.$